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RESEARCH ARTICLE

3D-Printed Demultiplexer Circuits Using Suspended-in-Air Grating Couplers for Terahertz Communications

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ABSTRACT Integrated photonic circuits are in great demand for the upcoming THz communications. This work explores 3D printing to realize high-quality, high-refractive-index-contrast integrated components and devices for demultiplexing terahertz channels within the Wavelength Division Multiplexing modality. Namely, by printing integrated circuits using Polypropylene filaments suspended in air, we profit from the high-refractive index contrast of such a material combination to realize relatively compact low-loss waveguides, bends, couplers, and fiber Bragg gratings. The two-nozzle FDM printer allows simultaneous printing with filaments of two distinct sizes of 800um and 400um, with the larger filament used to make waveguides and couplers, and the smaller one used to define high-quality fiber Bragg gratings containing as much as 100 periods and featuring stop bands as wide as 10 GHz. Furthermore, by employing judiciously designed mechanical supports we show how to integrate such subcomponents into functional components such as single-channel drop filters. Finally, we developed a low-loss splicing technique for joining several components into functional devices and demonstrated four-channel THz WDM demultiplexers with in-plane (horizontal) and a more compact out-of-plane (vertical) integration. Experimentally, three-channel demultiplexers of THz signals with individual data rates up to 6 Gbps were demonstrated. Using finite element numerical modeling, integrated circuits were optimized for operation in the 120-165 GHz frequency band featuring ~5 GHz individual channel bandwidths and ~3 GHz inter-channel spectral spacing, and good agreement with the experiments was observed. Additionally, the measured spectra closely resemble the simulated ones but exhibit a frequency shift of several GHz towards higher frequencies. Experimental results further reveal strong sidelobe suppression and a broader Drop port bandwidth (~6 GHz vs. ~4 GHz predicted). However, the measured Drop amplitudes (~0.5–0.6) are lower than the theoretical predictions (~0.8) due to ~10% scattering losses per supporting structure. We believe that the suspended-in-air integrated terahertz circuits hold strong potential for developing various linear optic transformers that will play a key role in energy-efficient analog processing of data streams for the upcoming terahertz communications. This is because of the high quality of the resultant circuits, ease of fabrication, and low infrastructure costs necessary for their manufacturing, thus allowing low-cost fast turnaround prototyping and development of terahertz signal processing devices even with the simplest 3D printing systems.

INDEX TERMS 3D printing, demultiplexer, integrated photonic circuits, wavelength-division multiplexer (WDM).

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I. INTRODUCTION

Most wireless systems currently operate in the overcrowded microwave band, which is insufficient to meet the future

bandwidth demand. Shifting the carrier wave to higher frequencies is essential to accommodate the anticipated surge in data volume [1], [2]. Consequently, the terahertz (THz) frequency band (0.1-10 THz) is regarded as the next frontier for wireless communication systems [3], [4], [5], [6]. Terahertz (THz) waves situated between microwave and infrared spectral bands attracted much interest for a variety of industrial applications in sensing [7], [8], imaging [9], [10], and security [11], [12] due to their many unique properties. Specifically for communications [3], [4], [13], [14], [15], [16], [17], [18], [19], THz waves enable larger bandwidths than microwaves, potentially enabling several 100 Gbps data rates per channel without any multiplexing. The expected increase in data traffic over the next decade, driven by nascent technologies such as the Internet of Things, Virtual and Augmented Reality, Artificial Intelligence, Big Data, etc. has prompted the development of the sixth generation (6G) wireless networks that greatly surpass existing 4G and 5G network standards.

Terahertz communications is a promising technology for the 6G networks capable of terabit-per-second wireless and fiber-assisted transmission for various data-demanding applications [20], [21], [22], [23], [24], [25], [26], [27]. To date, several demonstrations of free-space ultra-high bit rate data transfer (>100 Gbps) employed single-channel THz links with optical multiplexing [28], [29], [30], as well as advanced modulation techniques including quadrature amplitude modulation and quadrature phase-shift keying [31], [32], [33]. One promising way of data multiplexing in the THz range is Frequency Division Multiplexing (FDM), which encodes different channels using distinct carrier frequencies [34], [35]. A multiplexer (Mux) and a demultiplexer (Demux) are essential components in FDM technology. The multiplexer combines light from spatially separated spectrally distinct sources into a single data stream, while the demultiplexer spatially separates the multiplexed frequencies into single carrier frequency channels [36], [37], [38], [39].

The fabrication of THz devices often involves significant complexity. A standard way uses the methods and infrastructure of silicon photonics to fabricate low-loss THz optical elements [40], [41], [42], [43]. While scalable for mass production, this approach requires access to very expensive infrastructure and entails high running costs, which makes it ill-suited for rapid prototyping. Alternatively, certain thermoplastics used in additive manufacturing exhibit transparency to terahertz (THz) radiation, which opens an interesting opportunity for 3D printing as a cost-efficient alternative for rapid prototyping of THz devices. As a result, 3D printing has recently gained considerable attention within the Terahertz research community, leading to the development of numerous 3D-printed THz components such as freeform microwave waveguides, antennas, and basic optical elements [44], [45], [46], [47], [48], [49], [50], [51]. Furthermore, 3D printing opens a way for high-density integration of photonic circuits in three dimensions, a feat challenging to

accomplish with other methods. For example, Ortiz-Martinez et al. developed a 3D-printed filter made of polystyrene (PS) for the sub-THz 200-300 GHz band [52]. Weidenbach et al. demonstrated 3D-printed waveguide designs, including low-loss splitters and couplers, operating at 120 GHz, fabricated from polystyrene [53]. Additionally, 3D-printed terahertz grating couplers have been designed and characterized at 120 GHz for outcoupling and focusing THz radiation [54]. Furthermore, low-loss, low-dispersion waveguides printed from PS have achieved error-free performance at a data rate of 1 Gb/s [55].

Compared to recent studies, our work presents a novel 3D-printed terahertz demultiplexer featuring suspended-in-air grating couplers, offering significant advantages in fabrication flexibility, cost-effectiveness, and scalability. Unlike the photonic crystal-based demultiplexers proposed by Li et al. [56], which rely on cascaded directional coupling waveguides created by selectively removing rows of silicon rods, our approach eliminates the need for complex lithographic fabrication and precise photonic crystal alignment. Furthermore, while Wu et al. [57] demonstrated a metamaterial-based demultiplexer with high isolation and low insertion loss, its resonance-dependent design requires high precision fabrication and limits both scalability and tunability. In contrast, our 3D-printed demultiplexer circuits allow for customized integration into various THz systems, providing a more adaptable and practical solution for terahertz communication applications. Our presented demultiplexer enables precise, wavelength-specific separation of THz signals while offering significant advantages in terms of fabrication flexibility, cost-effectiveness, and scalability. Our method leverages the versatility of even the basic FDM 3D printing to create complex geometries that are difficult to achieve with conventional techniques, ultimately enhancing signal separation and overall system performance.

In this work, we demonstrate experimentally 3-channel demultiplexers with 2D (in-plane) and 3D (out-of-plane) integration for Frequency Division Multiplexing in the 100-200 GHz spectral band using Side-Coupled Waveguide Bragg Grating Filters as enabling building blocks. The devices were fabricated from a low-loss Polypropylene dielectric using Fused Filament Fabrication (Raise 3D Pro2 series). To the best of our knowledge, this is the first time that devices of such complexity have been realized using 3D printing, in the planar and vertically integrated variants. The advanced optical performance of our demultiplexers stems from the use of polypropylene (PP) polymer in air- a high refractive index contrast material combination that offers one of the lowest absorption losses in the THz regime. However, the 3D printing process introduces several challenges, including warping, adhesion difficulties, and variations in nozzle speed, printed waveguide size, and filament quality, all of which can affect the precision and performance of the printed gratings. To mitigate these limitations and ensure high fabrication accuracy, we have systematically optimized the

printing parameters, effectively minimizing structural deviations (judged from microscopy images) and enhancing the overall reliability of the fabricated demultiplexers (judged by repeatability of the performance from sample to sample).

The paper is organized as follows. First, we discuss the numerical design and optimization of Side-Coupled Waveguide Bragg Grating Filters capable of dropping individual 3 GHz-wide channels with center frequencies in the 140-170 GHz range, while letting other frequencies pass through. Particular attention is paid to the design and fabrication of Waveguide Bragg Gratings [58], [59] which are principal enabling elements of our spectral filters. Then, we discuss the spectral characterization of several such filters using an in-house photonics-based THz communication system [14]. Finally, we show how such filters can be integrated into 3-channel demultiplexers using either in-plane (2D) or out-of-plane (3D) integration strategies and conduct spectral and Bit Error Rate characterization of the resultant devices.

II. DESIGN OF THE WDM THZ FILTERS

Recent advancements in integrated waveguide Bragg grating (WBG) devices have highlighted their potential in microwave and IR for applications such as optical filtering, tunable delays, optical differentiation, and single-sideband modulation. WBG devices offer compactness, flexibility, and high efficiency, making them ideal for applications that require high-performance signal processing. Their ability to operate at THz bandwidths and integrate with other photonic components opens new possibilities for the development of high-speed, low-cost microwave photonic systems [60], [61]. Oh et al. demonstrated a tunable wavelength filter using Bragg gratings in polymer waveguides, leveraging the thermo-optic effect to efficiently shift the Bragg reflection wavelength. This approach achieved a tuning range of over 10 nm with low insertion loss [62].

The Side-Coupled Waveguide Bragg Grating Filter used in our work is an integrated photonic device for filtering out a specific wavelength of light. The schematic of such a device is shown in Fig. 1. (a). The principle of operation of such a device is as follows. The THz light with a mix of wavelengths (e.g. λ_1 , λ_2 , λ_3) enters the filter through the In port, propagates through a bend, and encounters a waveguide Bragg grating. Wavelengths that fall outside of the grating stopband (e.g. λ_2 , λ_3) continue through the grating while being redirected from the top waveguide to the bottom waveguide by the directional coupler, finally exiting the Through port. At the same time, wavelengths falling into the Bragg grating stopband (e.g. λ_1) are reflected by the grating while being redirected from the top waveguide into the bottom waveguide by the directional coupler, finally exiting at the Drop port.

The design process for such a filter involves three main steps: designing the Waveguide Bragg Grating, designing the directional coupler, and optimizing the overall structure. The first step is to design the Waveguide Bragg Grating, which is responsible for dropping a specific wavelength.

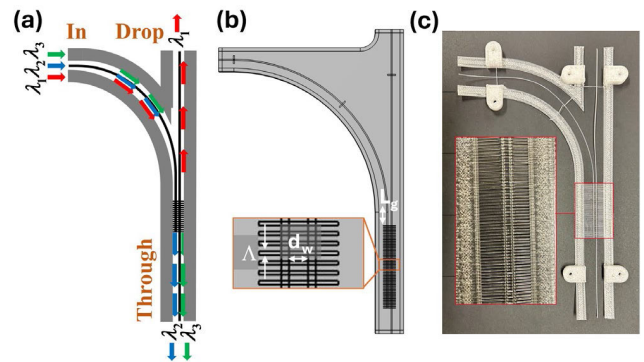


FIGURE 1. (a) A schematic of a WDM demultiplexer (b) A typical computational cell used in numerical simulations. (c) A photo of a WDM demultiplexer. Inset: zoom of a grating section.

This involves selecting a grating period (Λ) based on the desired drop frequency. The period is chosen to ensure that the Bragg grating stop band is centered around one of the channel carrier frequencies (e.g. 140 GHz). The second step involves designing a directional coupler, which includes a circular arc, and two waveguides running directly under the Bragg Grating. First, a separation distance d_w between two parallel waveguides is set to result in a relatively short device that can be printed using a 30 cm x 30 cm build plate of a 3D printer. Smaller values of d_w result in stronger coupling between two waveguides, and, thus, smaller device size. Next, the number of grating periods N_g is estimated to achieve near-zero transmission for wavelengths within the grating stop band. The exact number of periods in the grating is chosen, so that a second channel outside of the grating stop band (e.g. 145 GHz) is diverted from the launch waveguide into the Through port of a coupler. Finally, the standoff distance between the bend termination and the grating L_g is chosen so that the intensity in the Drop port is maximized at the center frequency of a dropped channel. To perform optimizations, we used finite element COMSOL Multiphysics software with a typical computation cell shown in Fig. 1 (b).

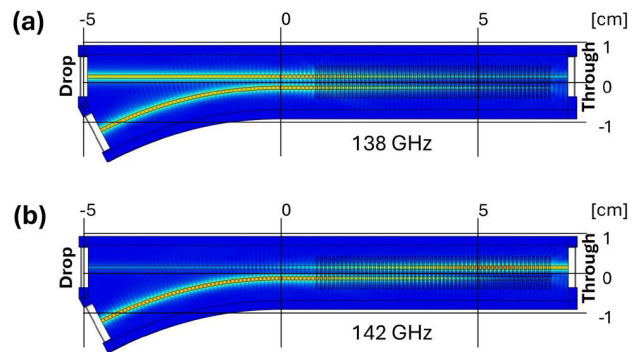


FIGURE 2. Electric field amplitude distribution in the Demux1 operating at (a) Drop (138 GHz) and (b) Through (142 GHz) frequencies.

Three devices were numerically optimized to have Drop channels centered around 138 GHz, 142 GHz, and 146 GHz,

with the Through frequencies ~ 4 GHz above the Drop channel center frequencies. In Fig. 2 electric field amplitude distributions are shown for Demux1 operating at the Drop (138 GHz) (panel a)) and Through (142 GHz) frequencies (panel b)). Drop and Through transmission coefficients (by power) are shown in Fig. 4 (a) and (b) (computed using numerical cells shown in Fig. 2). Finally, the geometric parameters of three multiplexers including the grating period, the number of grating periods, the distance between two parallel waveguides, and the standoff distance between the bend termination and the grating are summarized in Table 1.

TABLE 1. Geometrical parameters of three demultiplexers.

Demultiplexer	Demux1	Demux2	Demux3
Drop channel design center frequency (GHz)	138	142	146
Grating period Λ (μm)	1020	984	949
Number of grating periods	57	63	76
Inter-waveguide distance d_w in a coupler (mm)	1.95	1.95	1.95
A standoff distance L_g between the bend termination and the grating (mm)	9	15	21

III. EXPERIMENTAL REALIZATION OF THE WDM THZ FILTERS

The experimental realization of thus designed filters is challenging. The difficulty arises from the need for precise alignment of the waveguide Bragg grating relative to the waveguide coupler and providing mechanical support for the suspended-in-air components of the structure. To this end, the filter components are printed within a rigid hollow frame that features several slender support elements. The first printed layer contains slender support attached to the frame. The second layer contains all the waveguides printed on top of the supports using a 0.8 mm diameter nozzle. Finally, the third layer contains Bragg grating printed on top of the waveguide layer using a 0.4 mm diameter nozzle. The grating extends to the support frame for mechanical stability. To splice several filters together a 1 mm-diameter glass capillary of 100 μm wall thickness is used to align and put in contact the Through and In waveguides of the two filters. The capillary is then heated to fuse the plastic waveguides and then removed by shattering. Finally, the device is assembled on the optical bench with the frame kept under light tension using alignment screws. A photo of a typical demultiplexer is shown in Fig. 1 (c).

IV. OPTICAL CHARACTERIZATION OF THE WDM THZ FILTERS

Optical characterization of the 3D-printed demultiplexers was conducted using an in-house photonics-based THz communication system detailed earlier [46]. The schematic of the optical characterization setup is presented in Fig. 3(a), while Fig. 3(c) shows a photo of the measurement setup with a mounted device. Briefly, in the transmitter arm, two DFB lasers, independently tunable and operating within the infrared C-band with somewhat mismatched center frequencies, are combined using a 3 dB coupler and sent to a fiber-coupled photomixer to generate THz waves of fixed frequency anywhere in the ~ 0.1 -1 THz range with bandwidth of ~ 10 MHz. In the THz CW spectroscopy mode: the THz radiation of a set frequency from the photomixer (Model: IOD-PMD-14001 from NTT Electronics Inc) is guided through a WR-6 rectangular waveguide flange [see Fig. 3(a)], which is butt-coupled to the device under study. On the receiving end, a 10.8 mm diameter horn antenna collects the THz waves, which are then detected using a zero-bias Schottky detector (Model: WR8.0 ZBD-F from Virginia Diodes Inc). A high-gain, low-noise amplifier (Model: SLNA-030-32-30-SMA from Fairview Microwave Inc) is then used to amplify the received signal for further processing. In the THz communication mode [see Fig. 3 (b)]: a baseband signal source, generated by a pulse pattern generator integrated into the test equipment, produces pseudorandom bit sequences with varying bit rates. This signal undergoes amplification and modulation utilizing RF and Mach-Zehnder modulation techniques. The modulated laser beams are further amplified and injected into a photomixer to generate a modulated THz carrier wave. In the receiver section, the THz carrier wave is detected and demodulated using a zero-bias Schottky diode, then amplified through a low-noise amplifier. Eye pattern and bit error rate (BER) are then recorded. In more details, at the emitter side, the combined infrared optical signal from the coupler is modulated using an external electro-optic modulator (Models: LN81S-FC and MX10A from Thorlabs Inc). The modulated optical signal, with a fixed output power, is then amplified using an Erbium-Doped Fiber Amplifier (EDFA). At the receiver side, a Bias-Tee filters the DC field from the demodulated baseband signal, and a low-noise amplifier (LNA) further amplifies the received signal. Finally, the demodulated baseband signals are analyzed using a high-speed oscilloscope and a BER tester (Model: MP2100B from Anritsu Corporation). The BER measurements were conducted by varying the bit rate from 1 Gbps to 6 Gbps or adjusting the carrier frequency at a fixed bit rate. At each bit rate or carrier frequency, the decision threshold was optimized to balance insertion errors (digital zero misidentified as one) and omission errors (digital one misidentified as zero), thereby minimizing the BER.

In the experiments, the THz carrier wave in the 120 – 165 GHz spectral range was generated using an

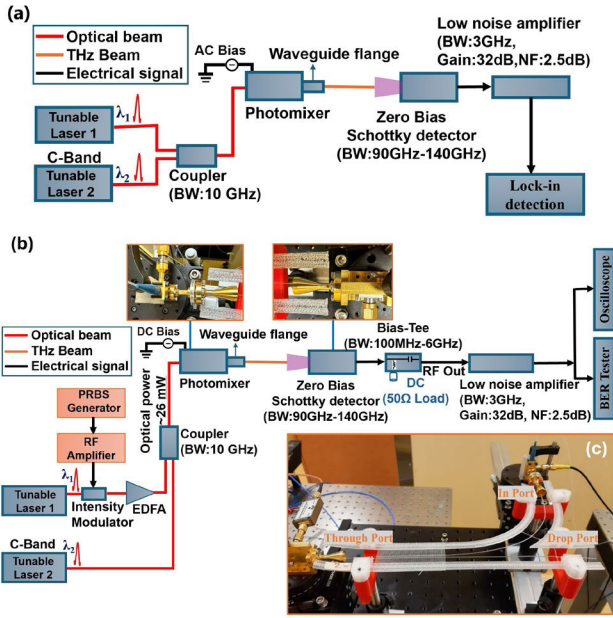


FIGURE 3. (a) Schematic of the continuous wave THz spectroscopy system. (b) Schematic of the photonics-based THz communication system. (c) Photo of the measurement setup with a mounted device.

optical photomixer (IOD-PMD-14001, NTT Electronics) with a photocurrent of 7mA generating THz powers between $125 \mu\text{W}$ (-9 dBm) and $250 \mu\text{W}$ (-6 dBm), and then coupled into a demultiplexer via a 1-inch-long WR6.5 rectangular waveguide terminated with a horn (WR8.0 ZBD-F, Virginia Diodes). A similar configuration was utilized at the output port of a demultiplexer where a horn was connected via a 1-inch-long WR6.5 rectangular waveguide to the Schottky diode. Operation of the Drop and Through ports of the filters were then characterized using two complimentary measurement modes, namely, THz CW spectroscopy and BER characterization for data transmission.

First, CW THz spectroscopy was conducted on three demultiplexers without data modulation. To extract the relative Drop and Trough coefficients (shown in Figs. 4(c,d)) for direct comparison with theoretical predictions (computed using numerical cells shown in Fig. 2) we normalize the raw transmission data by the transmission of a stand-alone bent waveguide ($R_b = 10 \text{ cm}$, 90° -bend) identical to those used in the filters (see insert in Fig. 4(c)). We note that overall, numerically computed spectra (shown in Figs. 4(a,b)) have very similar shapes to the measured ones. Additionally, as per the design goal, the higher-frequency edge of the Trough spectra shows higher transmission than the lower-frequency edge. At the same time, the experimental spectra are shifted to higher frequencies by several GHz. Moreover, we observe strong sidelobe suppression outside of the stopband in experimental spectra which results in the $\sim 6 \text{ GHz}$ experimental bandwidths at the Drop port, which is somewhat larger than numerically predicted bandwidths of $\sim 4 \text{ GHz}$. Finally, the maximal amplitudes of the measured

Drop spectra are somewhat lower (~ 0.5 - 0.6) compared to the theoretical ones (~ 0.8), due to scattering losses on supporting structures (10% scattering loss per structure predicted numerically using numerical cell shown in Fig. 1(b)), as well as due to nonuniformities of 3D-printed waveguides including wall roughness and micro-bending. The overall insertion loss is estimated to be 1.7 dB for the Drop port, and 1 dB for the Through port, which implies that with even a modest power budget of 11 dB, one can build a 10-channel THz WDM system using such filters placed in sequence.

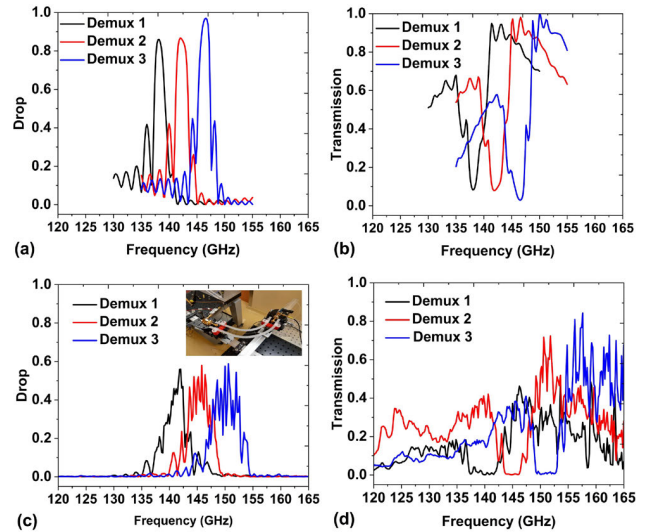


FIGURE 4. (a) Drop and (b) Through spectra as predicted by numerical simulations. Relative (c) Drop and (d) Through coefficients when using transmission through a stand-alone bend as a reference. Insert: picture of a stand-alone waveguide bend.

Next, we characterize information transmission through the 3 thus-developed multiplexors. Using Fig. 4, we chose the channel center frequencies as follows: demultiplexer 1 (Drop 140 GHz, Through 145 GHz), demultiplexer 2 (Drop 145 GHz, Through 150 GHz), and demultiplexer 3 (Drop 150 GHz, Through 155 GHz). Therefore, all the following BER measurements will be conducted at those frequencies. Specifically, the eye patterns for the THz data streams of various bit rates ranging from 1 to 6 Gbps using amplitude-shift-keying modulation were recorded, and corresponding Bit Error Rate (BER) measurements were conducted. During BER measurements, the decision threshold was adjusted to balance the insertion error (incorrectly identifying a digital 0 as a digital 1) and omission error (incorrectly identifying a digital 1 as a digital 0). The duration of recording was determined as $1/(\text{target BER} \times \text{bit rate})$, with the target BER set to 10^{-12} (error-free transmission threshold). First, the performance of Demux1 was characterized in the Drop mode at the carrier frequency corresponding to the grating stopband center frequency of 140 GHz. Next, the performance of Demux1 was characterized in the Through mode using the carrier outside of the grating stopband at 145 GHz as shown in Fig. 5(a). Similarly, performances of Demux2 and Demux3

were characterized in Drop mode at the corresponding stop-band center frequencies of 145 GHz and 150 GHz, as well as in Through mode outside of the corresponding stopbands at 150 GHz and 155 GHz as shown in Figs. 5(b,c). Experimental data confirms that all demultiplexers can operate with 6 Gbps data streams with $\text{BER} < 10^{-4}$ below the forward error correction limit of $\sim 10^{-3}$.

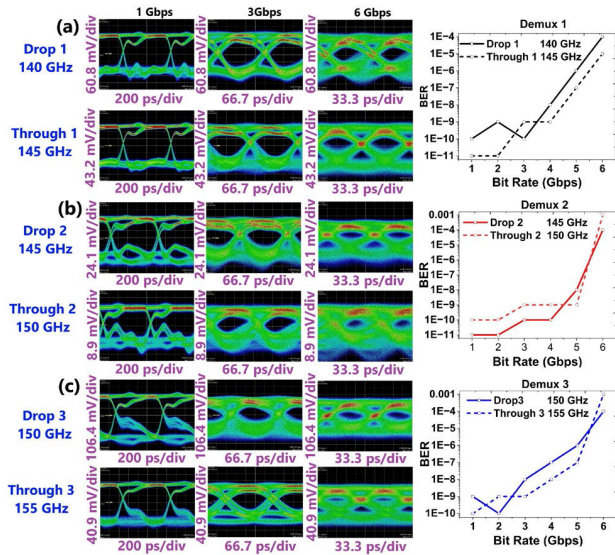


FIGURE 5. Measured Bit Error Rate (BER) versus Bit Rate in the 1-6 Gbps range and the corresponding eye diagrams for 3 demultiplexers (a) Demux1 (b) Demux2, and (c) Demux3 at the Drop and Through ports.

V. WDM DEVICES – TWO DEMULTIPLEXERS CONNECTED IN SEQUENCE

In the following, we demonstrate 3-channel demultiplexers using two filters from the previous section connected in sequence (schematically shown in Fig. 6(a)). The first device was made by splicing the Through 1 and In 2 ports of Demux 1 and Demux 2 (Demux 1+2), while the second one was made by splicing the Through 2 and In 3 ports of Demux 2 with Demux 3 (Demux 2+3) as shown in Fig. 6(b). Then, spectroscopic and BER measurements were performed at various Drop and Through ports.

Fig. 7 (a) shows the Drop 1 spectrum, while Fig. 7 (b) shows Drop 2/Through 2 spectra in the range of 120–165 GHz for Demux 1+2. Similarly, Fig. 7 (c) shows the Drop 2 spectrum, while Fig. 8 (d) shows Drop 3 and Through 3 spectra for Demux 1+3.

The communication performance of in-sequence demultiplexers was evaluated by measuring the bit error rate (BER) while adjusting the bit rate from 1 to 6 Gbps. Specifically, for Demux 1+2, BER was measured at 140 GHz for the Drop 1 port, 145 GHz for the Drop 2, and 150 GHz for the Through 2 ports. The measured bit error rate (BER) versus bit rate, along with the corresponding eye diagrams for Demux 1+2 are shown in Figure 8 (a-d). Similarly, for Demux 2+3 BER was measured at 145 GHz for the Drop 1 port, 150 GHz

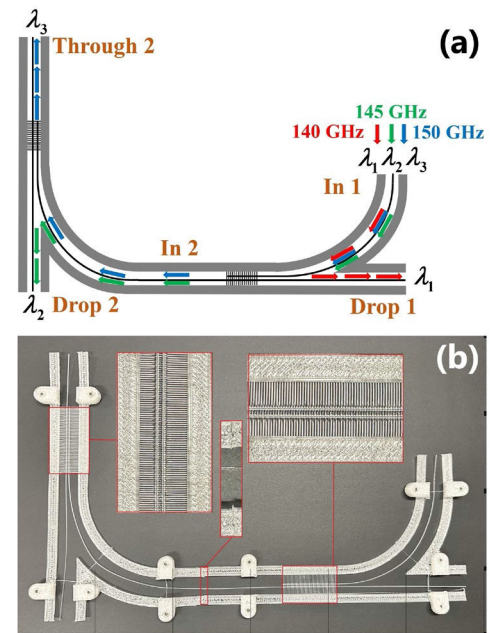


FIGURE 6. (a) A schematic of the in-sequence demultiplexer (Demux 1+2). (b) A photo of the assembled device.

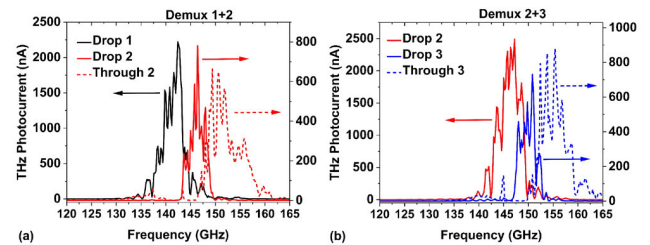


FIGURE 7. Experimental spectra of (a) Drop 1, (b) Drop 2, and Through 2 of the Demux 1+2. Experimental spectra of (c) Drop 2, (d) Drop 3, and Through 3 of the Demux 2+3.

for the Drop 3 port, and 155 GHz for the Through 3 port. The measured BER versus bit rate, along with the corresponding eye diagrams, are presented in Figure 8 (d-h).

VI. WDM DEVICES – TWO DEMULTIPLEXERS CONNECTED IN PARALLEL

Finally, we demonstrate 3-channel demultiplexers using two filters connected in parallel (schematically shown in Fig. 9(a)). The first device was made by splicing the In and Through ports of Demux 1 and Demux 2 (Demux 1||2), while the second one was made by splicing the In and Through ports of Demux 2 with Demux 3 (Demux 2||3) as shown in Fig. 9(b). Then, spectroscopic and BER measurements were performed at various Drop and Through ports.

Experimental spectra at the Drop 1, Drop 2, and Through 1+2 ports of the in-parallel Demux 1||2 are shown in Fig. 10 (a) in the spectral range of 120–165 GHz. Similarly, experimental spectra at the Drop 2, Drop 3, and Through 2+3 ports of the in-parallel Demux 2||3 are shown in Fig. 10 (b).

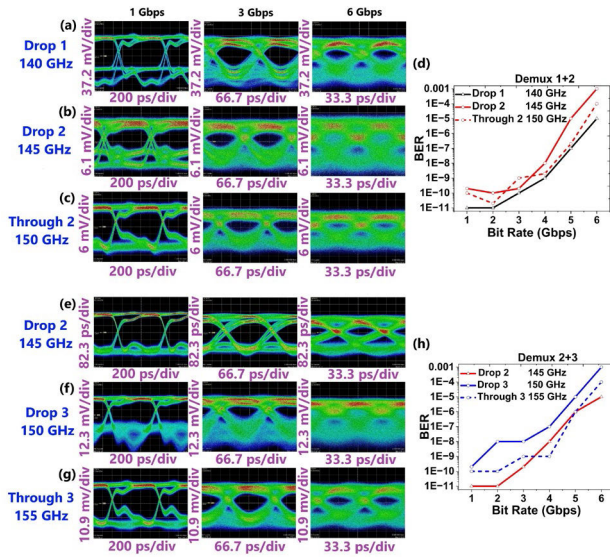


FIGURE 8. Measured Bit Error Rate versus Bit Rate and the corresponding eye diagrams for in-sequence Demux 1+2 at different bitrates (a) eye pattern at the Drop 1 port (b) eye pattern at the Drop 2 port (c) eye pattern at the Through 2 port (d) BER at the Drop 1, Drop 2, and Through 2 ports. Similar data is shown in panels (e)-(h) for Demux 2+3 and ports Drop 2, Drop 3, and Through 3 ports.

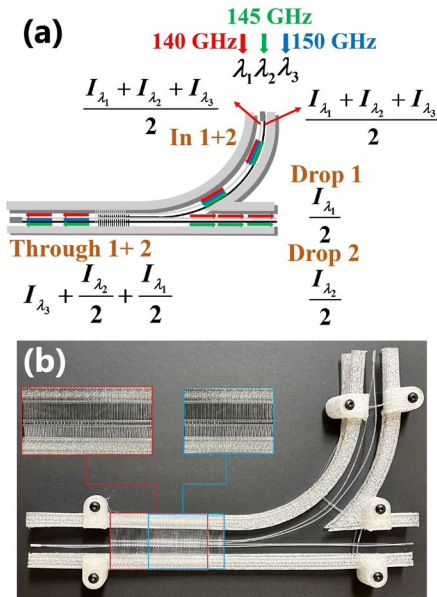


FIGURE 9. (a) A schematic of the in-parallel demultiplexer (Demux 1||2). (b) A photo of the assembled device.

The communication performance of the in-parallel Demux 1||2 was evaluated by measuring the bit error rate (BER) for bit rates between 1 and 6 Gbps. Specifically, BER was measured at 140 GHz for Drop 1, 145 GHz for Drop 2, and 140 GHz, 145 GHz, and 150 GHz for the composite Through 1+2 ports. The measured bit error rate (BER) versus bit rate, along with the corresponding eye diagrams for the Demux 2|3 are shown in Figs. 11 (a-d). Similarly, for Demux 2|3

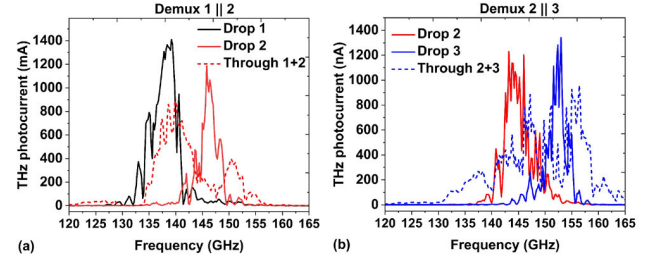


FIGURE 10. (a) Experimental Drop 1, Drop 2, and through 1+2 spectra of the Demux 1||2. (b) Experimental Drop 2, Drop 3, and through 2+3 spectra of the Demux 2||3.

BER was measured at 145 GHz for Drop 1, 150 GHz for Drop 3, and 145 GHz, 150 GHz, and 155 GHz for Through 2+3 ports. The measured BER versus bit rate, along with the corresponding eye diagrams, are presented in Figs. 11 (e-h).

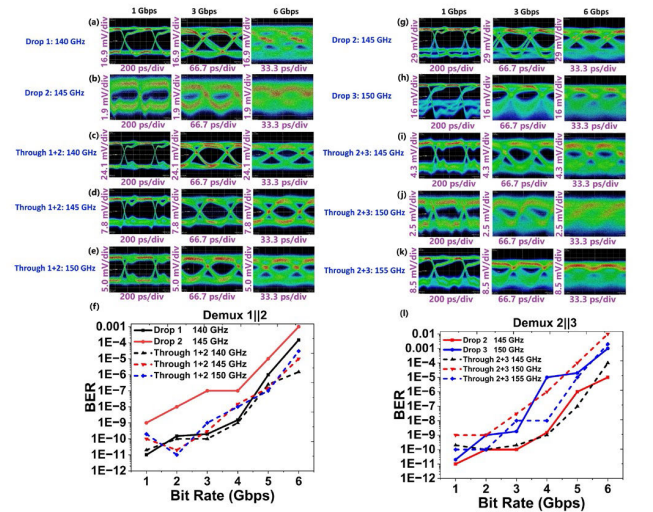


FIGURE 11. Measured bit error rate versus bit rate and the corresponding eye diagrams for in-parallel Demux 1||2 at different bitrates. Eye patterns at the (a) Drop 1 port, (b) Drop 2 port, (c,d,e) through 1+2 port. (e) BER at the Drop 1, Drop 2, and through 1+2 ports. Similar data is shown in panels (g)-(i) for Demux 2||3 and ports Drop 2, Drop 3, and through 2+3.

VII. CONCLUSION

In this work, three THz demultiplexer filters, as well as their combinations in sequence and in parallel were fabricated and characterized for operation with four sub-mm wave channels (140 GHz, 145 GHz, 150 GHz, and 155 GHz) within the WDM framework of Terahertz Communication. Successful channel demultiplexing of 3 channels per device was demonstrated with up to 6Gbps data rates. Experimental results further demonstrate strong sidelobe suppression and a broader Drop port bandwidth (~6 GHz compared to the predicted ~4 GHz). However, the measured Drop amplitudes (~0.5–0.6) are lower than the theoretical predictions (~0.8), primarily due to approximately 10% scattering losses per supporting structure. Our study indicates that additive manufacturing of THz circuits with a

high-refractive-index-contrast Polypropylene-in-air material combination presents a viable approach for high-quality low-loss fabrication of integrated THz devices. Additionally, additive manufacturing carries a strong potential for three-dimensional integration of THz circuits, as well as a fast turn-around between design and prototyping.

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